# ECE 6390 Satellite Communication and Navigation Systems Fall 2011 Homework Assignment 2

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#### Assignment

In class, we worked through an analytical example using acoustic diffraction theory to find the radiation pattern due to a circular aperture of radius R with uniform field illumination. In this homework problem, you will explore how the patterns of dish antennas change as the illumination changes. Solution will require a relatively straight-forward numerical evaluation of the area integration we derived in class. You may use any software package or technique available to you (be sure to attach your code/inputs).

## 1 Solution

Using the Kirchoff-Helmholtz Integral Theorem for far-field evaluation:

$$\tilde{E}_{\circ}(r,\theta,\varphi) = \frac{jk\cos(\theta)}{4\pi r} e^{-jkr} \int_{0}^{R} \int_{0}^{2\pi} E_{\circ}(\rho') e^{jk\sin(\theta)\rho'\cos(\varphi')} d\varphi'\rho' d\rho'$$

The electric field illumination that we will use across the circular aperture will take the form: (-1)

$$E_{\circ}(\rho') = E_o \cos^n\left(\frac{\pi \rho'}{R}\right), for |\rho| \le R,$$

where  $E_o$  is an arbitrary constant and n is the unitless "aperture taper factor" (n=0, of course, corresponds to uniform illumination).

$$\tilde{E}_{\circ}(r,\theta,\varphi) = \frac{jkcos(\theta)}{4\pi r} e^{-jkr} \int_{0}^{R} \int_{0}^{2\pi} E_{o} \cos^{n}\left(\frac{\pi\rho'}{R}\right) e^{jksin(\theta)\rho'\cos(\varphi')} d\varphi'\rho'd\rho' =$$
  
Note:  $J_{n}(x) = \frac{1}{2\pi} \int_{-\pi}^{\pi} e^{\left(-j(n\tau - xsin\tau)\right)} d\tau$ 

We evaluate the integral and obtain the  $0^{\text{th}}$  order Bessel function:

$$=\frac{jkcos(\theta)}{4\pi r}e^{-jkr}E_o\int_0^R\cos^n\left(\frac{\pi\rho'}{R}\right)2\pi J_0(ksin(\theta)\rho')\rho'd\rho'=$$

Note: The wave number,  $k = \frac{2\pi}{\lambda}$ 

$$= \frac{j\pi \cos(\theta)}{\lambda r} e^{-jkr} E_o \int_0^R \cos^n\left(\frac{\pi\rho'}{R}\right) J_0\left(\frac{2\pi\sin(\theta)\rho'}{\lambda}\right) p' d\rho' =$$
  
Directivity =  $D(\theta, \varphi) = \frac{\|\vec{s}\|}{\|\vec{s}_{iso}\|} = \frac{|\vec{E}^2|}{|\vec{E}^2_{iso}|} = \frac{4r^2|\vec{E}^2|}{E_0R^2}$   
 $\|\vec{S}\| = \frac{1}{2}Re\{ExH^*\} = \frac{\hat{z}|E_0^2|}{2\eta_0}$ 

$$D(\theta,\varphi) = \frac{k^2 \cos^2 \theta}{R^2} \left[ \int_0^R \cos^n \left( \frac{\pi \rho'}{R} \right) J_0(k \sin \theta \rho') p' d\rho' \right]^2$$

### 1.1 Simulation

```
% ECE 6390 HW #2 Code
% 27 October 2011
% Modified by Andrew Punnoose and Mason Nixon
% from code by Prof. Durgin
clear all
close all
clc
for n = 0:3 % aperture taper factor
  R = 5; % dish radius in wavelengths (unitless)
  M = 500; % number of points to plot
  d = .01; % numerical integration step size
  Gmin = -40; % minimum plotting level
  theta = linspace(-pi/2,pi/2,M);
  rhop = 0:d:R;
  D = [];
```

```
for th = theta
        func = (\cos(pi.*rhop./R).^n).*besselj(0,2.*pi.*sin(th).*rhop).*rhop;
        drhop = rhop(2) - rhop(1);
        integral = sum(func).*drhop;
        D = [D, integral];
    end
    Dir = (4.*pi.^2.*(abs(D).^2).*(cos(theta).^2))./(R.^2);
    DdB = 10 \times \log 10 (Dir);
    DdB = max(DdB - max(DdB) - Gmin, 0);
    figure();
    polar(theta+pi/2, DdB);
    t = sprintf('Radiation pattern for n=%1.0f',n);
    title(t)
    Gpeak = max(DdB);
                                      % peak gain computation
    hpbw = sum(DdB(1:M) >= Gpeak/2)*pi/M; % calculate half-power BW
    ind = DdB(2:(M/2))>DdB(1:(M/2-1));
                                                    % find local maxima in
gain
    SLL = max(DdB(([0 ind] - [ind 1])>0))/Gpeak; % ... and use for SLL
    % Display results
    fprintf('\n\n Uniform dish of %1.1f wavelengths', R);
    fprintf('\n -----');
    fprintf('\n Peak gain:
                                        %3.1f (%2.1f dBi)', ...
        Gpeak, 10*log10(Gpeak) );
    fprintf('\n Half-power beadwidth: %2.1f deg', hpbw*180/pi );
fprintf('\n Side-lobe level: %2.1f dB', -10*log10(SLL) );
    fprintf('\n\n');
end
```

# 2 Results

The following results were obtained for a radius of  $R=5\lambda$  and aperture taper factors of n=0,1,2,3 respectively:



Peak gain: 40.0 (16.0 dBi) Half-power beadwidth: 17.3 deg Side-lobe level: 2.5 dB



Peak gain: 40.0 (16.0 dBi)

Half-power beadwidth: 31.0 deg Side-lobe level:  $0.5~\mathrm{dB}$ 



Peak gain: 40.0 (16.0 dBi) Half-power beadwidth: 28.8 deg Side-lobe level: 1.7 dB



Peak gain: 40.0 (16.0 dBi) Half-power beadwidth: 38.2 deg Side-lobe level: 0.6 dB The following results were obtained for a radius of  $R=10\lambda$  and aperture taper factors of n=0,1,2,3 respectively:



Peak gain: 40.0 (16.0 dBi) Half-power beadwidth: 9.4 deg Side-lobe level: 2.5 dB



Peak gain: 40.0 (16.0 dBi) Half-power beadwidth: 15.8 deg Side-lobe level: 0.4 dB



Peak gain: 40.0 (16.0 dBi) Half-power beadwidth: 14.4 deg Side-lobe level: 1.6 dB



Peak gain: 40.0 (16.0 dBi) Half-power beadwidth: 20.2 deg Side-lobe level: 0.5 dB

The results show that the odd aperture taper factors result in a larger half-power beamwidth and larger side levels than in the even aperture taper factors. Additionally, for an aperture radius of  $10\lambda$ , we see a narrower main lobe, indicating a more focused beam. This seems intuitive since as we increase aperture size, we get a more focused beam pattern.